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# *Study of Microwave theory*

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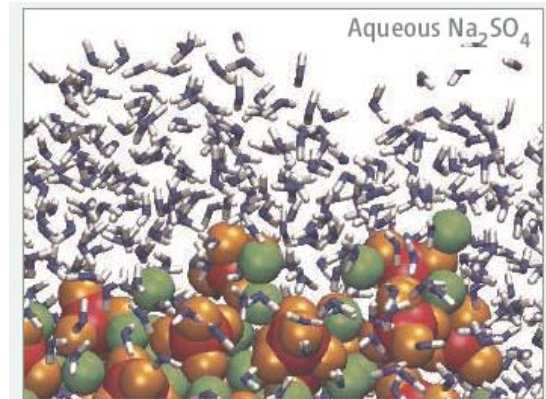
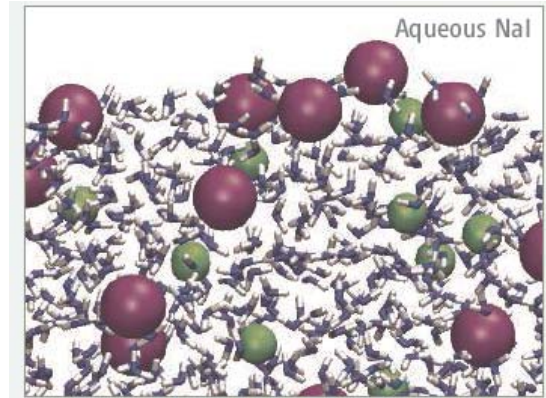
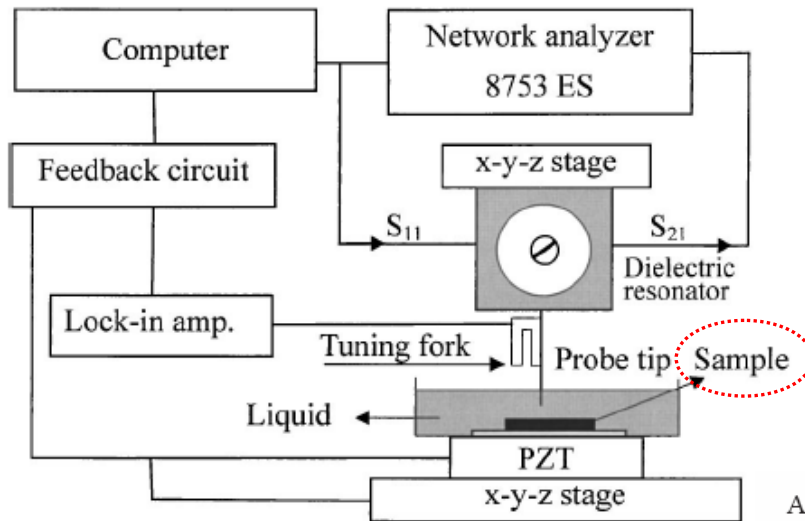


# Introduction

- Is there **surface enhancement of ionic salts** in aqueous solution ?



- Can we **measure** ?



APPLIED PHYSICS LETTERS 86, 153506 (2005)

# Introduction

APPLIED PHYSICS LETTERS 89, 183504 (2006)

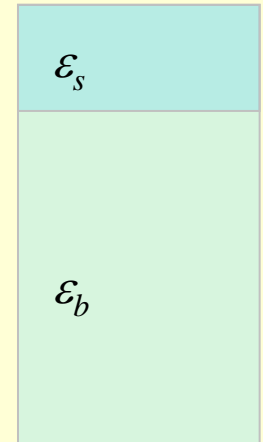


$$S_{11} = 20 \log \left| \frac{Z^R - Z_0}{Z^R + Z_0} \right|$$



$$Z^R = \frac{Z_a}{\sqrt{\epsilon_s}} \frac{1 + [\tan(k_a(v/s) \sqrt{\epsilon'_0 + c\gamma'})]^2}{1 + [(\epsilon'_0 + c\gamma')/\epsilon_s][\tan(k_a(v/s) \sqrt{\epsilon'_0 + c\gamma'})]^2}$$

two layer model



$\epsilon_{sample}$

$Z_0$  : effective impedance of the probe tip

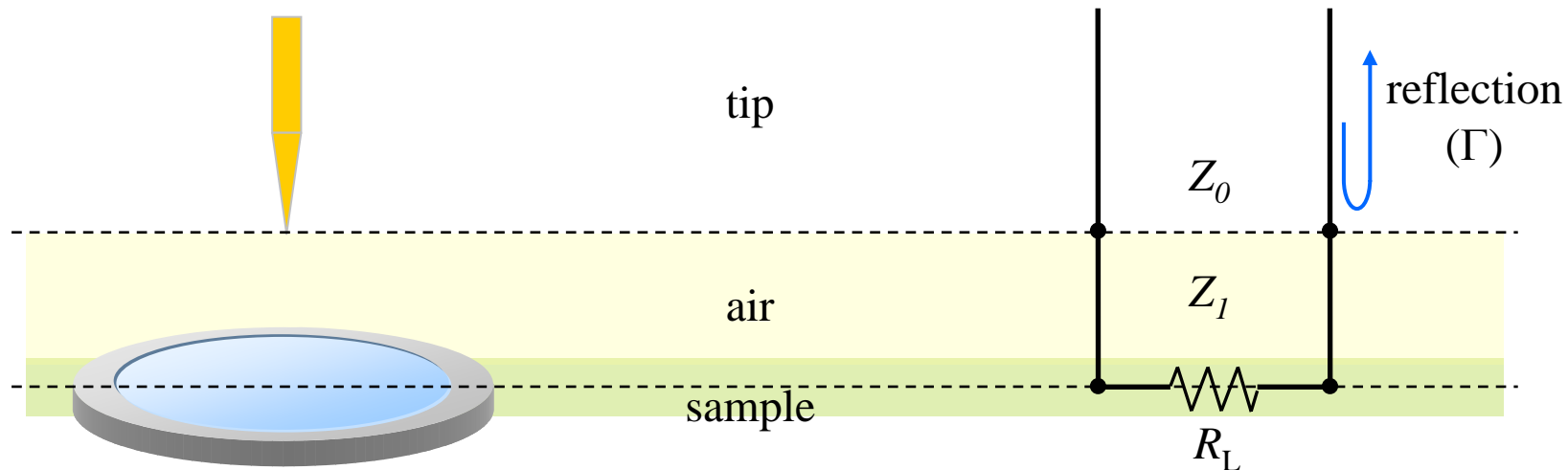
$Z^R$  : real part of the complex impedance of solution

$k_a$  : wavenumber in air

$v$  : volume of solution

$s$  : surface area of solution

# Our goal - matching with the circuit



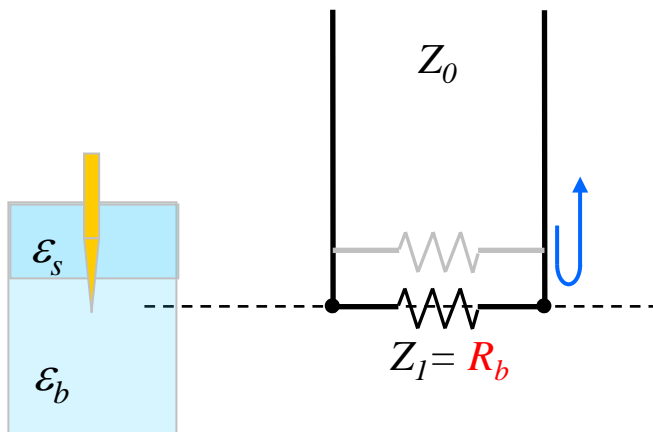
$$Z_1 = \sqrt{Z_0 R_L}$$



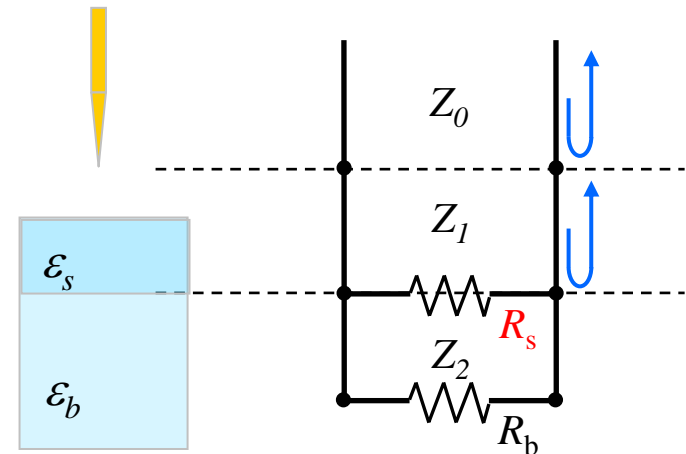
$$|\Gamma| = \left| \frac{Z_{in} - Z_0}{Z_{in} + Z_0} \right|$$

# Outline

(1) first step




(2) second step



# Wave equation - in lossless medium

$$\begin{cases} \vec{\nabla} \times \vec{E} = -j\omega\mu\vec{H} \\ \vec{\nabla} \times \vec{H} = j\omega\varepsilon\vec{E} \end{cases} \quad \Rightarrow \quad \begin{cases} \vec{\nabla}^2 \vec{E} + \omega^2 \mu\varepsilon \vec{E} = 0 \\ \vec{\nabla}^2 \vec{H} + \omega^2 \mu\varepsilon \vec{H} = 0 \end{cases}$$

$= k^2$ , where  $k = \omega\sqrt{\mu\varepsilon}$

sol. 

$$E_x(z) = E^+ e^{-jkz} + E^- e^{jkz}$$

or  $E_x(z) = E^+ \cos(\omega t - kz) + E^- \cos(\omega t + kz)$

$$H_y(z) = \frac{1}{\eta} (E^+ e^{-jkz} - E^- e^{jkz})$$

where  $\eta = \mu\omega/k = \sqrt{\mu/\varepsilon}$  ; wave impedance for the plane wave  
in free space,  $\eta_0 = \sqrt{\mu_0/\varepsilon_0} = 377 \Omega$


# Wave equation - in general lossy medium

(conductive medium)

$$\begin{cases} \vec{\nabla} \times \vec{E} = -j\omega\mu\vec{H} \\ \vec{\nabla} \times \vec{H} = j\omega\varepsilon\vec{E} + \sigma\vec{E} \end{cases} \implies \vec{\nabla}^2 \vec{E} + \omega^2 \mu\varepsilon \left( 1 - j \frac{\sigma}{\omega\varepsilon} \right) \vec{E} = 0$$

$= -\gamma^2$ , where

$$\gamma \equiv \alpha + j\beta = j\omega\sqrt{\mu\varepsilon} \sqrt{1 - j \frac{\sigma}{\omega\varepsilon}}$$

sol.  


$$E_x(z) = E^+ e^{-\gamma z} + E^- e^{\gamma z}$$

$$e^{-\gamma z} = e^{-\alpha z} e^{-j\beta z}$$

$$\rightarrow e^{-\alpha z} \cos(\omega t - \beta z)$$

decay is given by the attenuation constant,  $\alpha$

\* phase velocity  $v_p = \omega / \beta$ , a wavelength  $\lambda = 2\pi / \beta$

\* if lossless :  $\alpha = 0$ ,  $\beta = k$ ,  $\gamma = jk$

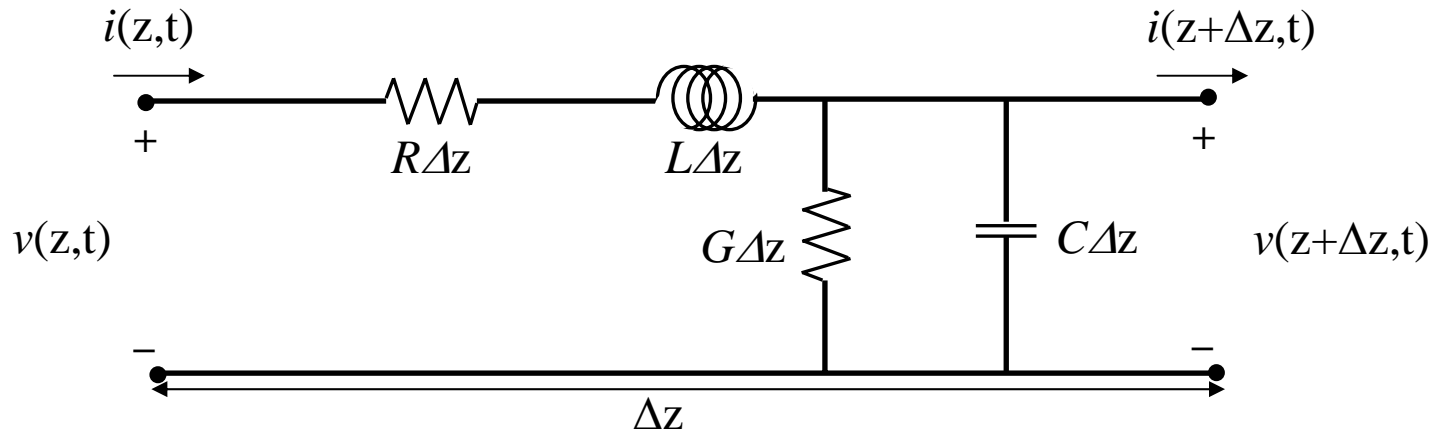
# Summary of wave propagation - in various media

	type of Medium		
	lossless ( $\epsilon'' = \sigma = 0$ )	general lossy	good conductor ( $\epsilon'' \gg \epsilon'$ or $\sigma \gg \omega\epsilon'$ )
complex propagation constant	$\gamma = j\omega\sqrt{\mu\epsilon}$	$\gamma = j\omega\sqrt{\mu\epsilon}$ $= j\omega\sqrt{\mu\epsilon'}\sqrt{(1 - j\sigma)/\omega\epsilon'}$	$\gamma = (1 + j)\sqrt{\omega\mu\sigma/2}$
phase constant (wavenumber)	$\beta = k = \omega\sqrt{\mu\epsilon}$	$\beta = \text{Im}(\gamma)$	$\beta = \text{Im}(\gamma) = \sqrt{\omega\mu\sigma/2}$
attenuation constant	$\alpha = 0$	$\alpha = \text{Re}(\gamma)$	$\alpha = \text{Re}(\gamma) = \sqrt{\omega\mu\sigma/2}$
impedance	$\eta = \mu\omega/k = \sqrt{\mu/\epsilon}$	$\eta = j\mu\omega/\gamma$	$\eta = (1 + j)\sqrt{\mu\omega/2\sigma}$
skin depth	$\delta_s = \infty$	$\delta_s = 1/\alpha$	$\delta_s = \sqrt{2/\mu\omega\sigma}$
wavelength	$\lambda = 2\pi/\beta$	$\lambda = 2\pi/\beta$	$\lambda = 2\pi/\beta$
phase velocity	$v_p = \omega/\beta$	$v_p = \omega/\beta$	$v_p = \omega/\beta$

\*  $\epsilon = \epsilon' - j\epsilon''$



# Transmission Line Theory



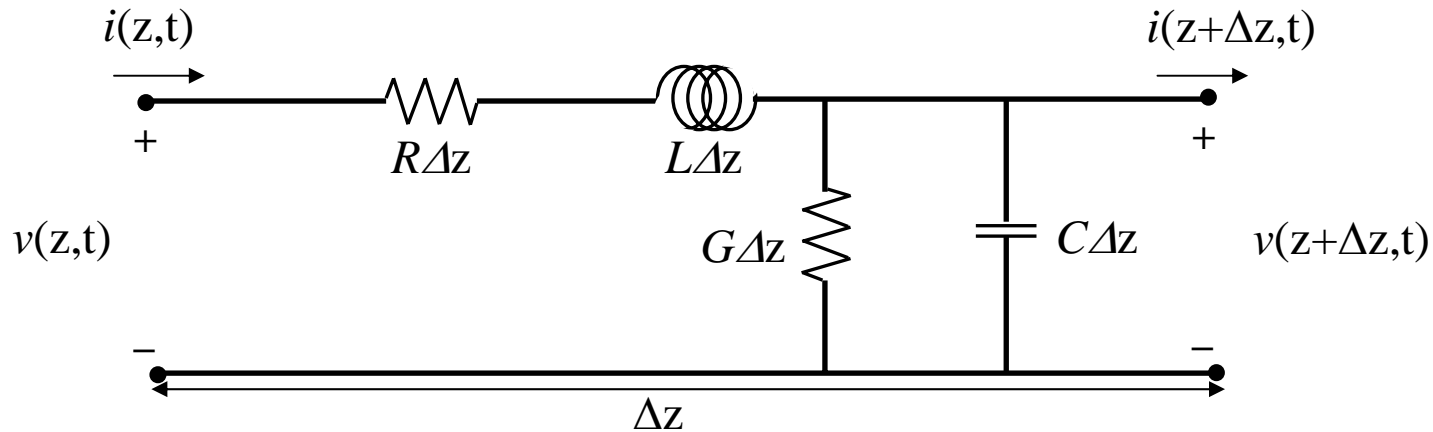
$R$  = series resistance per unit length length, for both conductors, in  $\Omega/\text{m}$ .  
 $L$  = series inductance per unit length length, for both conductors, in  $\text{H}/\text{m}$ .  
 $G$  = shunt conductance per unit length length, in  $\text{S}/\text{m}$ . (\* $G = 1/R$ )  
 $C$  = series inductance per unit length length, in  $\text{F}/\text{m}$ .

by the Kirchhoff's law

$$v(z,t) - R\Delta z \cdot i(z,t) - L\Delta z \frac{\partial i(z,t)}{\partial t} - v(z + \Delta z,t) = 0$$

$$i(z,t) - G\Delta z \cdot v(z + \Delta z,t) - C\Delta z \frac{\partial v(z + \Delta z,t)}{\partial t} - i(z + \Delta z,t) = 0$$

# Transmission Line Theory



$$\begin{cases} v(z,t) - R\Delta z \cdot i(z,t) - L\Delta z \frac{\partial i(z,t)}{\partial t} - v(z + \Delta z,t) = 0 \\ i(z,t) - G\Delta z \cdot v(z + \Delta z,t) - C\Delta z \frac{\partial v(z + \Delta z,t)}{\partial t} - i(z + \Delta z,t) = 0 \end{cases}$$

taking the limit as  $\Delta z \rightarrow 0$  gives the following

$$\begin{cases} \frac{\partial v(z,t)}{\partial z} = -Ri(z,t) - L \frac{\partial i(z,t)}{\partial t} \\ \frac{\partial i(z,t)}{\partial z} = -Gv(z,t) - C \frac{\partial v(z,t)}{\partial t} \end{cases}$$

# Transmission Line Theory

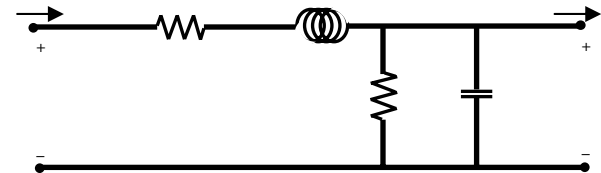
$$\left\{ \begin{array}{l} \frac{\partial v(z,t)}{\partial z} = -Ri(z,t) - L \frac{\partial i(z,t)}{\partial t} \\ \frac{\partial i(z,t)}{\partial z} = -Gv(z,t) - C \frac{\partial v(z,t)}{\partial t} \end{array} \right. \quad \begin{array}{l} i(z,t) = I(z)e^{j\omega t}, \\ v(z,t) = V(z)e^{j\omega t} \end{array} \quad \left\{ \begin{array}{l} \frac{dV(z)}{dz} = -(R + j\omega L)I(z) \\ \frac{dI(z)}{dz} = -(G + j\omega C)V(z) \end{array} \right.$$

$$\Rightarrow \left\{ \begin{array}{l} \frac{d^2 V(z)}{dz^2} = (R + j\omega L)(G + j\omega C)V(z) \\ \frac{d^2 I(z)}{dz^2} = (R + j\omega L)(G + j\omega C)I(z) \end{array} \right.$$

$\equiv \gamma$

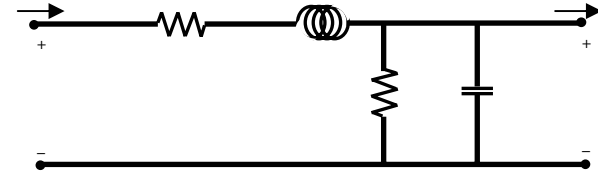
$$\Rightarrow \left\{ \begin{array}{l} \frac{d^2 V(z)}{dz^2} - \gamma^2 V(z) = 0 \\ \frac{d^2 I(z)}{dz^2} - \gamma^2 I(z) = 0 \end{array} \right.$$

where,  $\gamma = \alpha + j\beta = \sqrt{(R + j\omega L)(G + j\omega C)}$



# Transmission Line Theory

$$\begin{cases} \frac{d^2 V(z)}{dz^2} - \gamma^2 V(z) = 0 \\ \frac{d^2 I(z)}{dz^2} - \gamma^2 I(z) = 0 \end{cases}$$



sol.

$$\begin{cases} V(z) = V_0^+ e^{-\gamma z} + V_0^- e^{\gamma z} \\ I(z) = I_0^+ e^{-\gamma z} + I_0^- e^{\gamma z} \end{cases}$$

z direction    -z direction

$$\frac{dV(z)}{dz} = -(R + j\omega L)I(z)$$

$$\begin{aligned} \frac{dV(z)}{dz} &= \frac{d}{dz} (V_0^+ e^{-\gamma z} + V_0^- e^{\gamma z}) = -\gamma (V_0^+ e^{-\gamma z} - V_0^- e^{\gamma z}) \\ &= -(R + j\omega L)I(z) \end{aligned}$$

$$\therefore I(z) = \frac{\gamma}{R + j\omega L} (V_0^+ e^{-\gamma z} - V_0^- e^{\gamma z})$$

$\equiv 1/Z_0$

$z_0$  : characteristic impedance

$$= \frac{R + j\omega L}{\gamma} = \sqrt{\frac{R + j\omega L}{G + j\omega C}}$$

# Transmission Line Theory

$$I(z) = \frac{1}{Z_0} (V_0^+ e^{-\gamma z} - V_0^- e^{\gamma z}) = \frac{V_0^+ e^{-\gamma z}}{Z_0} - \frac{V_0^- e^{\gamma z}}{Z_0}$$

$$V(z) = V_0^+ e^{-\gamma z} + V_0^- e^{\gamma z}$$

$$\frac{V_0^+}{I_0^+} = Z_0 = -\frac{V_0^-}{I_0^-}$$

$$e^{-\gamma z} = e^{-\alpha z} e^{-j\beta z}$$
$$\rightarrow e^{-\alpha z} \cos(\omega t - \beta z)$$

converting back to the time domain,

$$v(z, t) = |V_0^+| \cos(\omega t - \beta z + \phi^+) e^{-\alpha z} + |V_0^-| \cos(\omega t + \beta z + \phi^-) e^{\alpha z}$$

# Lossless transmission Line ( $\alpha = 0$ )

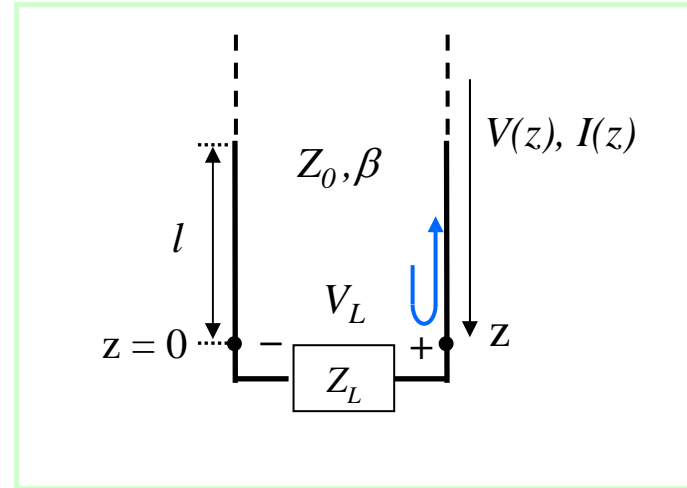
$$V(z) = V_0^+ e^{-j\beta z} + V_0^- e^{j\beta z}$$

$$I(z) = \frac{V_0^+}{Z_0} e^{-j\beta z} - \frac{V_0^-}{Z_0} e^{j\beta z}$$

$$Z_L = \frac{V(z=0)}{I(z=0)} = \frac{V_0^+ + V_0^-}{V_0^+ - V_0^-} Z_0$$

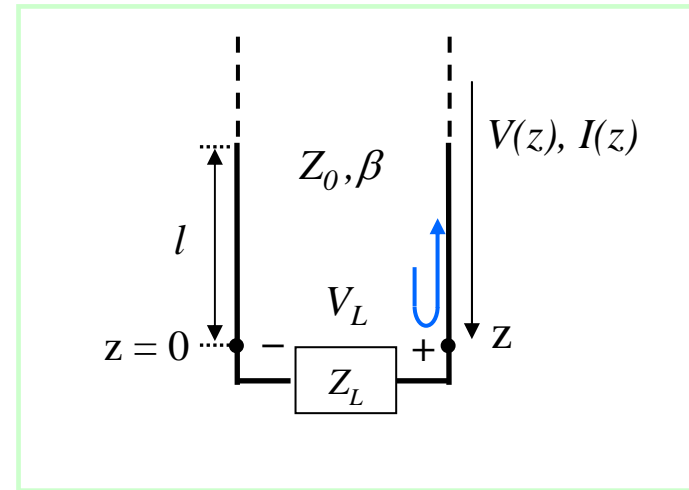
$$\Rightarrow V_0^- = \frac{Z_L - Z_0}{Z_L + Z_0} V_0^+$$

$$\therefore \Gamma = \frac{V_0^-}{V_0^+} = \frac{Z_L - Z_0}{Z_L + Z_0}$$



# Lossless transmission Line ( $\alpha = 0$ )

$$\Gamma(z=l) = \frac{V_0^- e^{-j\beta l}}{V_0^+ e^{j\beta l}} = \Gamma(0) e^{-2j\beta l}$$



$$Z_{in}(z=-l) = \frac{V(-l)}{I(-l)} = \frac{V_0^+ \left[ e^{j\beta l} + \Gamma e^{-j\beta l} \right]}{V_0^+ \left[ e^{j\beta l} - \Gamma e^{-j\beta l} \right]} Z_0 = \frac{1 + \Gamma e^{-2j\beta l}}{1 - \Gamma e^{-2j\beta l}}$$